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Short Bio

 Apr 2003 – Jun 2008: PhD Candidate, University of Patras
 Dissertation: "Statistical Characterization of Multipath Propagation and Received Signal in Time Varying Wireless Channels"

Dec 2009 – Jul 2016: University of Bedfordshire
 Research Fellow, Dec 2009 – Aug 2011
 Lecturer/Senior Lecturer, Sep 2011 – Jul 2016

>Aug 2016 – Dec 2021: Lecturer, University of Glasgow

≻Jan 2022 – Today: Associate Professor, Edinburgh Napier University



➤Wireless Channel Modeling

Wireless Communication System Model



➢<u>Wireless Channel Modeling (Small Scale)</u> Mathematically describe/characterize the (equivalent) wireless channel responses in compliance with the Maxwellian basis of electromagnetic wave propagation.

✓Absolutely necessary to design reliable/optimum wireless systems and components, i.e., antennas, multiplexers (OFDM, CDMA), diversity combiners!

Research Disciplines

Wireless Channel Modeling

Signal Processing

Statistics and Stochastic Processes Electromagnetic Wave Propagation Propagation Mechanisms

≻Line-of-sight (LOS)

≻Reflection and transmission

➢Diffraction

Scattering

➤Waveguiding

A. F. Molisch, "Wireless Communications," 2nd Edition, Wiley, 2011.

Propagation Mechanisms Line-of-sight (LOS)



Friis' formula:
$$P_r = P_t G_r G_t \left(\frac{\lambda_0}{4\pi d}\right)^2 = \frac{P_t G_r G_t}{L}$$

Free space loss :
$$L = \left(\frac{4\pi d}{\lambda_0}\right)^2$$



 $\varepsilon_{r1}, \varepsilon_{r2}$: Relative dielectric constants of materials 1 and 2, respectively $\theta_i, \theta_r, \theta_t$: Incidence angle, reflection angle, transmission angle, respectively \checkmark Snell's law: $\theta_r = \theta_i$ $\frac{\sin \theta_t}{\sin \theta_r} = \sqrt{\frac{\varepsilon_{r1}}{\varepsilon_{r2}}}$



Wave "bending" during the interaction with sharp edges
Causes power loss (shadowing), but makes propagation possible behind tall objects

▶<u>Rules-of-thumb</u>

Increased power loss with increased frequency (decreased wavelenght)
 Less evident as wavelength decreases with respect to sharpness size



✓ Results random number of arbitrarily directed diffuse components
✓ Smoothness and roughness are determined by the wavelength

<u>Rules-of-thumb</u>

Roughness dimension should be much larger than the wavelength
 Increased roughness with decreased wavelength (increased frequency)

Propagation Mechanisms Waveguiding

✓ Propagation through confined structures, e.g., street canyons, tunnels, corridors

Multipath Propagation

Multipath Propagation Vehicle-to-Vehicle (V2V) Communication Scenario

P. Karadimas and D. Matolak, "Generic Stochastic Modeling of Vehicle-to-Vehicle Wireless Channels," Elsevier Vehicular Communications vol. 1, no. 4, pp. 153-167, Oct. 2014.

Path Loss, Shadowing, Multipath

≻<u>Path loss</u>

✓Power decay due to transmission distance.

≻<u>Shadowing</u>

✓Power loss and large scale space-time variations due to obstruction by large objects.

><u>Multipath propagation</u>

✓Small scale space-time variations due to multipath components arrival and their interaction.

 $\log_{10} d$

Baseband Received Fading Signal

Wireless Channel Responses

$$H(\mathbf{\Omega},\tau,v) \xrightarrow{F_{v}^{-1}} F_{\mathbf{\Omega}}^{-1} \xrightarrow{h(\mathbf{r},\tau,t)} h(\mathbf{r},\tau,t) \xrightarrow{F_{\tau}} g(\mathbf{r},f,t)$$
$$H(\mathbf{\Omega},\tau,v) \xrightarrow{F_{v}^{-1}} F_{\mathbf{\Omega}}^{-1} \xrightarrow{F_{\tau}} g(\mathbf{r},f,t)$$

 $H(\Omega, \tau, v)$: directional-delay-Doppler variant channel response $h(\mathbf{r}, \tau, t)$: space-delay-time variant channel response $g(\mathbf{r}, f, t)$: space-frequency-time variant channel response

Tx – Rx Baseband Signal Relationships

$$y(\mathbf{r},t) = \iiint \exp[j2\pi(\mathbf{\Omega}\cdot\mathbf{r})/\lambda_0] \exp(j2\pi vt) x(t-\tau) H(\mathbf{\Omega},\tau,v) d\mathbf{\Omega} d\tau dv$$
(1)

$$y(\mathbf{r},t) = \int h(\mathbf{r},\tau,t) x(t-\tau) d\tau \quad (2)$$

$$y(\mathbf{r},t) = \int g(\mathbf{r},f,t) X(f) \exp(j2\pi ft) df \quad (3) \qquad x(t) \xleftarrow{F_t} X(f)$$

$$g(\mathbf{r}, f, t) = \sum_{l=1}^{L} a_l \exp[j2\pi(\mathbf{\Omega}_l \cdot \mathbf{r}) / \lambda_0] \exp(j2\pi v_l t) \exp(-j2\pi f \tau_l)$$
(4)

L: number of multipath components, λ_0 : carrier wavelength

Frequency Selective (Wideband) Modeling

J. D Parsons and A. S. Bajwa, "Wideband Characterization of Fading Mobile Radio Channels," IEE Proc. vol. 129, no. 2, pp. 95-101, Apr. 1982.

Narrowband Modeling

$$x(t - \tau_l) \approx x(t) \Leftrightarrow \tau_l \approx 0 \quad (5)$$

$$g(\mathbf{r}, f, t) = g(\mathbf{r}, t) = \sum_{l=1}^{L} a_l \exp[j2\pi(\mathbf{\Omega}_l \cdot \mathbf{r}) / \lambda_0] \exp(j2\pi v_l t) \quad (6)$$

 $y(\mathbf{r},t) = g(\mathbf{r},t)x(t) \quad (7)$

First – Order Statistical Modeling

(6):
$$g(\mathbf{r},t) = \sum_{l=1}^{L} |a_l| \exp(j\Phi_l) = A \exp(j\Phi)$$
 (7)

First – Order Statistical Modeling Probabilistic characterization of the amplitude A, i.e., the received signal envelope, and phase Φ in (7).

 ✓ Modeling should comply with the physical characteristics of the wireless propagation channel, e.g., Rayleigh distribution for A in the case of purely diffusive multipath propagation, Rice distribution in the case of a strong specular/LOS component added to purely diffusive multipath propagation. Second – Order Statistical Modeling

$$g(\mathbf{r},t) = \sum_{l=1}^{L} a_l \exp[j2\pi(\mathbf{\Omega}_l \cdot \mathbf{r}) / \lambda_0] \exp(j2\pi v_l t) \quad (6)$$

 $R(\Delta \mathbf{r}, \Delta t) = E[g^*(\mathbf{r}, t)g(\mathbf{r} + \Delta \mathbf{r}, t + \Delta t)] \quad (8)$

Second – Order Statistical Modeling
 Mathematical expansion of the expectation formula in
 (8) and further processing to characterize the space and time variability of the wireless propagation channel.

✓ Mathematical analysis should rely on the physical spatial characteristics of the wireless propagation channel, i.e., how multipath power is distributed in the three-dimensional space surrounding the Tx or the Rx.

Wireless Systems and Techniques Antenna Array Design* Cryptographic Key Generation**

*Y. Huang, P. Karadimas, and A.P. Sohrab, "Spatial Degrees of Freedom for Optimum Antenna Arrays," IEEE Trans. Wirel. Commun., under review, May 2022. **M. Bottarelli, P. Karadimas, G. Epiphaniou, D. K. B. Ismail and C. Maple, "Adaptive and Optimum Secret Key Establishment for Secure Vehicular Communications," IEEE Trans. Veh. Technol., vol. 70, no. 3, pp. 2310-2321, Mar. 2021.

Maximized Channel Capacities

Secret Bit Generation Rate – No Optimization

 α_{STD}

Optimized Secret Bit Generation Rate

Thank you for your attention

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